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# **RATE AND DETECTION ERROR-EXPONENT TRADEOFFS OF JOINT COMMUNICATION AND SENSING**

2nd IEEE International Hybrid Symposium on Joint Communications & Sensing



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▶ **GENERAL CONTEXT**

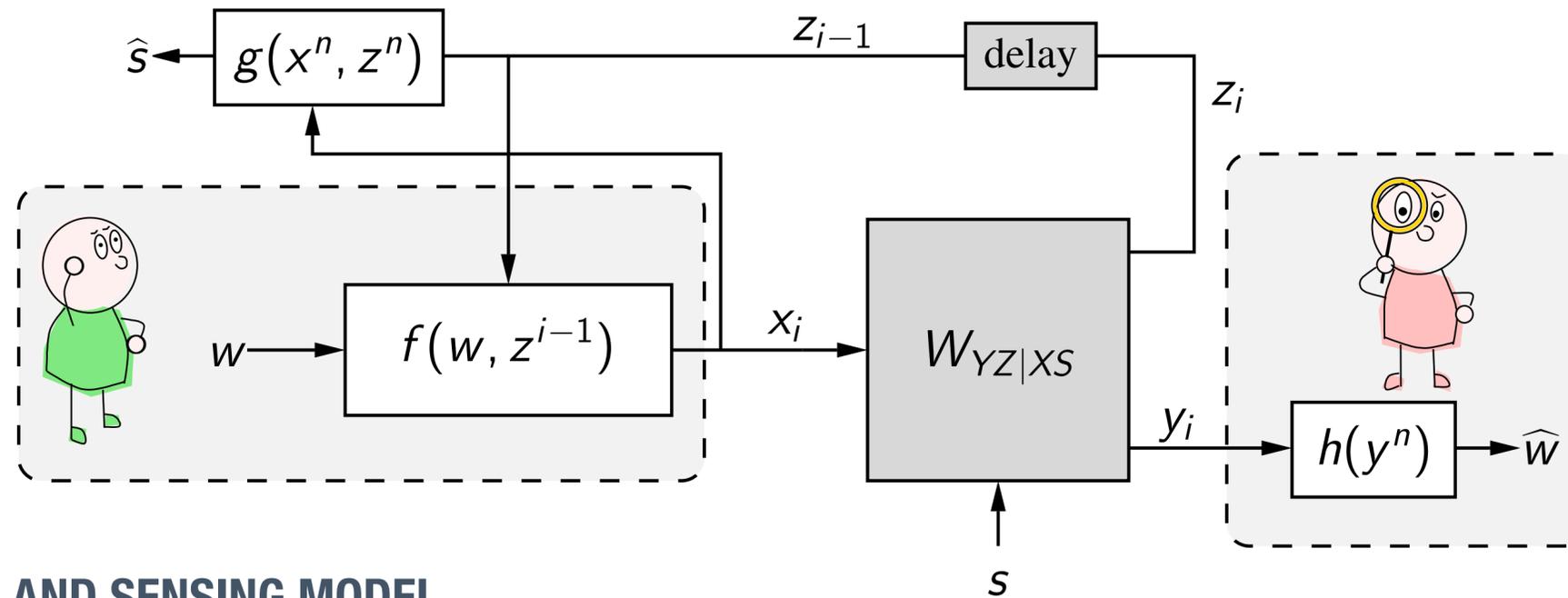
- ▶ mmWave systems enable convergence of radar and communication wavelengths
- ▶ Previously separated communication and sensing systems can coexist on a single hardware
- ▶ Sensing can significantly improve communication performance

▶ **INFORMATION-THEORETIC PERSPECTIVE ON JOINT COMMUNICATION AND SENSING**

- ▶ **Objective:** develop fundamental insights into benefits of joint approach
- ▶ **Previous works:** sensing independent and identically distributed (i.i.d.) channel state
  - ▶ [Zhang et al.'11, Kobayashi et al.'18'19, Ahmadipour et al.'21]
  - ▶ Characterization of rate-distortion region with (optimal) open-loop strategies

▶ **PRESENT WORK: MODEL WITH MEMORY**

- ▶ Discrete Memoryless Channel (DMC) with **fixed** channel state
  - ▶ **Motivation:** Channel change rate much slower than communication rate
- ▶ Full characterization of sensing vs. communication tradeoff in open loop
- ▶ Identification of significant benefits when **adapting** to channel with sensing

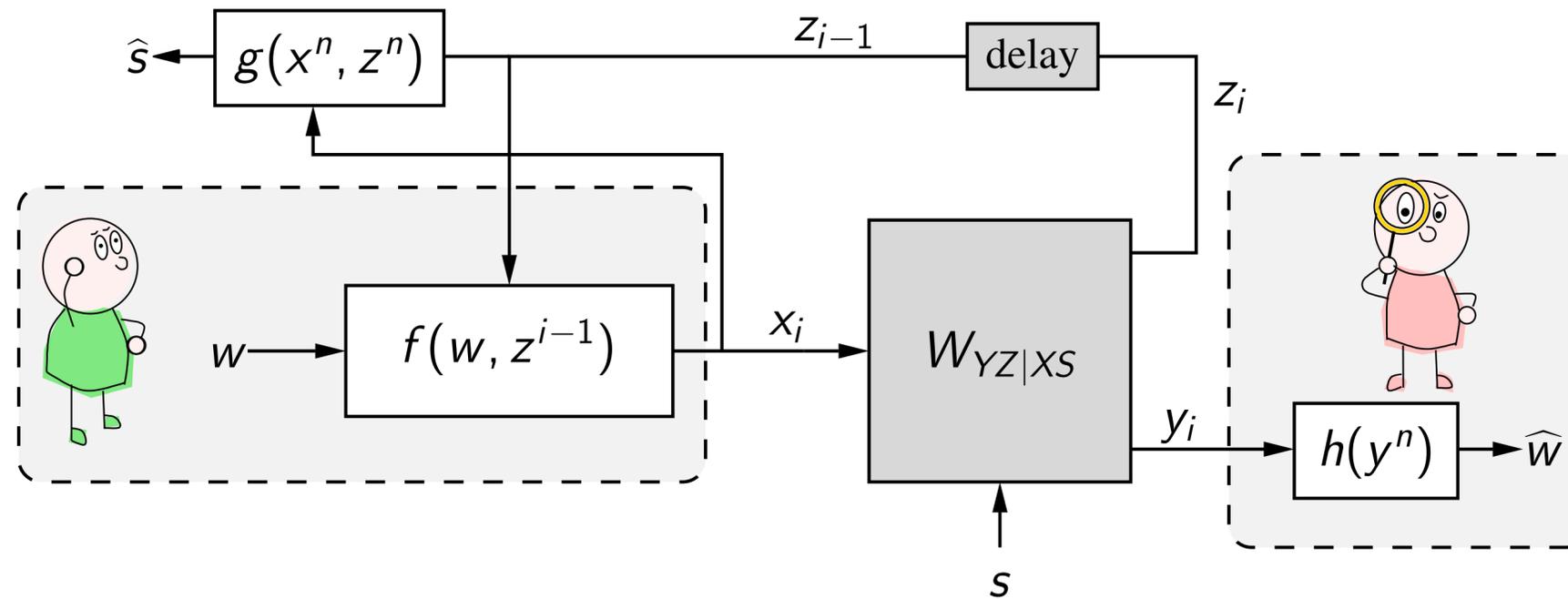


### ▶ JOINT COMMUNICATION AND SENSING MODEL

- ▶ State-dependent Discrete Memoryless Channel (Compound Channel)
- ▶ State  $s \in \mathcal{S}$ ,  $|\mathcal{S}| < \infty$
- ▶ Encoder:  $f_i : [1; M] \times \mathcal{Z}^{i-1} \rightarrow \mathcal{X} \forall i \in [1; n]$ , decoder:  $h : \mathcal{Y}^n \rightarrow [1; M]$ , estimator:  $g : \mathcal{X}^n \times \mathcal{Z}^n \rightarrow \mathcal{S}$

### ▶ PERFORMANCE METRICS

- ▶ Communication-error probability:  $P_c^{(n)} \triangleq \max_{s \in \mathcal{S}} \max_{w \in [1; M]} \mathbb{P}(h(Y^n) \neq w | W = w, S = s)$
- ▶ Detection-error probability:  $P_d^{(n)} \triangleq \max_{s \in \mathcal{S}} \frac{1}{M} \sum_{w=1}^M \mathbb{P}(g(Z^n) \neq s | S = s, W = w)$
- ▶ Rate:  $R \triangleq \frac{1}{n} \log M$  and detection error exponent  $E_d^{(n)} \triangleq -\frac{1}{n} \log P_d^{(n)}$



► **STRATEGIES: OPEN LOOP VS CLOSED LOOP**

- Open loop model does not use channel state information
- Closed loop model uses feedback for encoding
  - Adapt communication to sensing

► **EXISTENCE OF TRADEOFFS BETWEEN PERFORMANCE METRICS**

- No tradeoff between rate and estimation error: possible to transmit at capacity with vanishing estimation error
- Tradeoff is between rate and **detection error exponent**
- **Intuition:** both rate and detection error exponent depend on **type** of codewords

**THEOREM: RATE AND EXPONENT REGION FOR OPEN-LOOP**

Joint communication and sensing region

$$\mathcal{C}_{\text{open}} = \bigcup_{P_X \in \mathcal{P}_X} \left\{ \begin{array}{l} (R, E) \in \mathbb{R}_+^2 : \\ R \leq \min_{s \in \mathcal{S}} \mathbb{I}(P_X, W_{Y|X,s}) \\ E \leq \phi(P_X) \end{array} \right\}$$

where

$$\phi(P_X) = \min_{s \in \mathcal{S}} \min_{s' \neq s} \max_{\ell \in [0,1]} - \sum_x P_X(x) \log \left( \sum_z W_{Z|x,s}(z)^\ell W_{Z|x,s'}(z)^{1-\ell} \right)$$

**COROLLARY: NO TRADEOFF CONDITION**

If there exists  $x_0 \in \mathcal{X}$  such that for all  $x \in \mathcal{X}$  there exists a permutation  $\pi_x$  on  $\mathcal{Z}$  such that for every  $s \in \mathcal{S}$

$$W_{Z|X,s}(z|x) = W_{Z|X,s}(\pi_x(z)|x_0)$$

then

$$\mathcal{C}_{\text{open}} = \left\{ \begin{array}{l} (R, E) \in \mathbb{R}_+^2 : \\ R \leq \max_{P_X} \min_{s \in \mathcal{S}} \mathbb{I}(P_X, W_{Y|X,s}) \\ E \leq \max_{P_X} \phi(P_X) \end{array} \right\}$$

▶ **STATE-DEPENDENT PERFORMANCE METRICS**

- ▶ State-dependent communication-error probability:  $P_{c,s}^{(n)} \triangleq \max_w \mathbb{P}(h(Y^n) \neq w | W = w, S = s)$
- ▶ State-dependent detection-error probability:  $P_{d,s}^{(n)} \triangleq \max_w \mathbb{P}(h(Y^n) \neq w | W = w, S = s)$
- ▶ State-dependent detection-error exponent:  $E_{d,s}^{(n)} \triangleq -\frac{1}{n} \log P_{d,s}^{(n)}$

**THEOREM: INNER BOUND FOR RATE AND EXPONENT REGION FOR CLOSED-LOOP**

Joint communication and sensing region

$$\mathcal{C}_{\text{closed}}^s \supseteq \bigcup_{P_X \in \mathcal{P}_X} \left\{ \begin{array}{l} (R, E) \in \mathbb{R}_+^2 : \\ R \leq \mathbb{I}(P_X, W_{Y|X,s}) \\ E \leq \psi_s(P_X) \end{array} \right\}$$

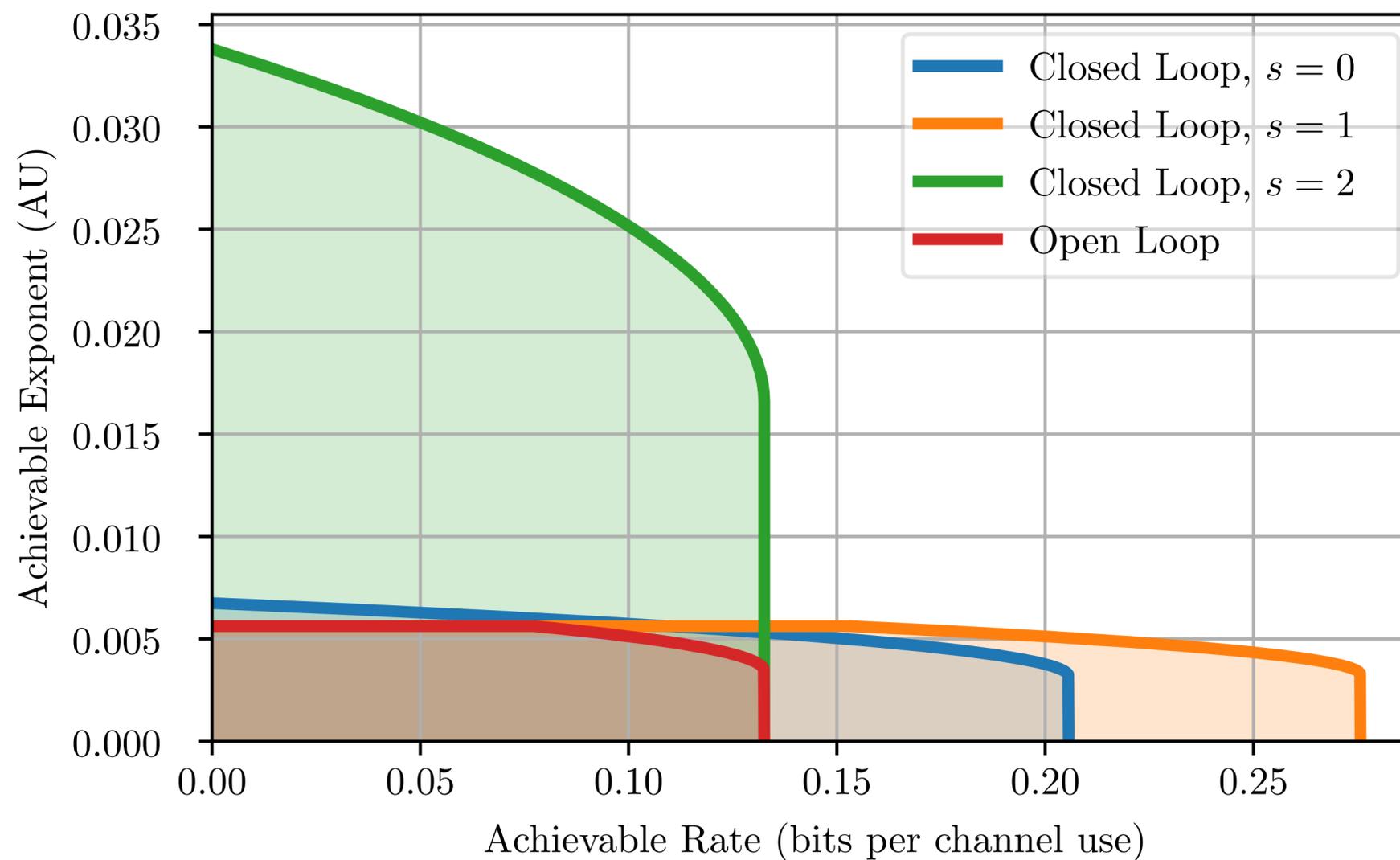
where

$$\psi_s(P_X) = \min_{s' \neq s} \max_{\ell \in [0,1]} - \sum_x P_X(x) \log \left( \sum_z W_{Z|X,s}(z|x)^\ell W_{Z|X,s'}(z|x)^{1-\ell} \right)$$

## ► SIMULATION RESULTS FOR OPEN LOOP AND CLOSED LOOP REGIONS

$S \backslash X$	0	1
0	0.9	0.3
1	0.8	0.2
2	0.7	0.2

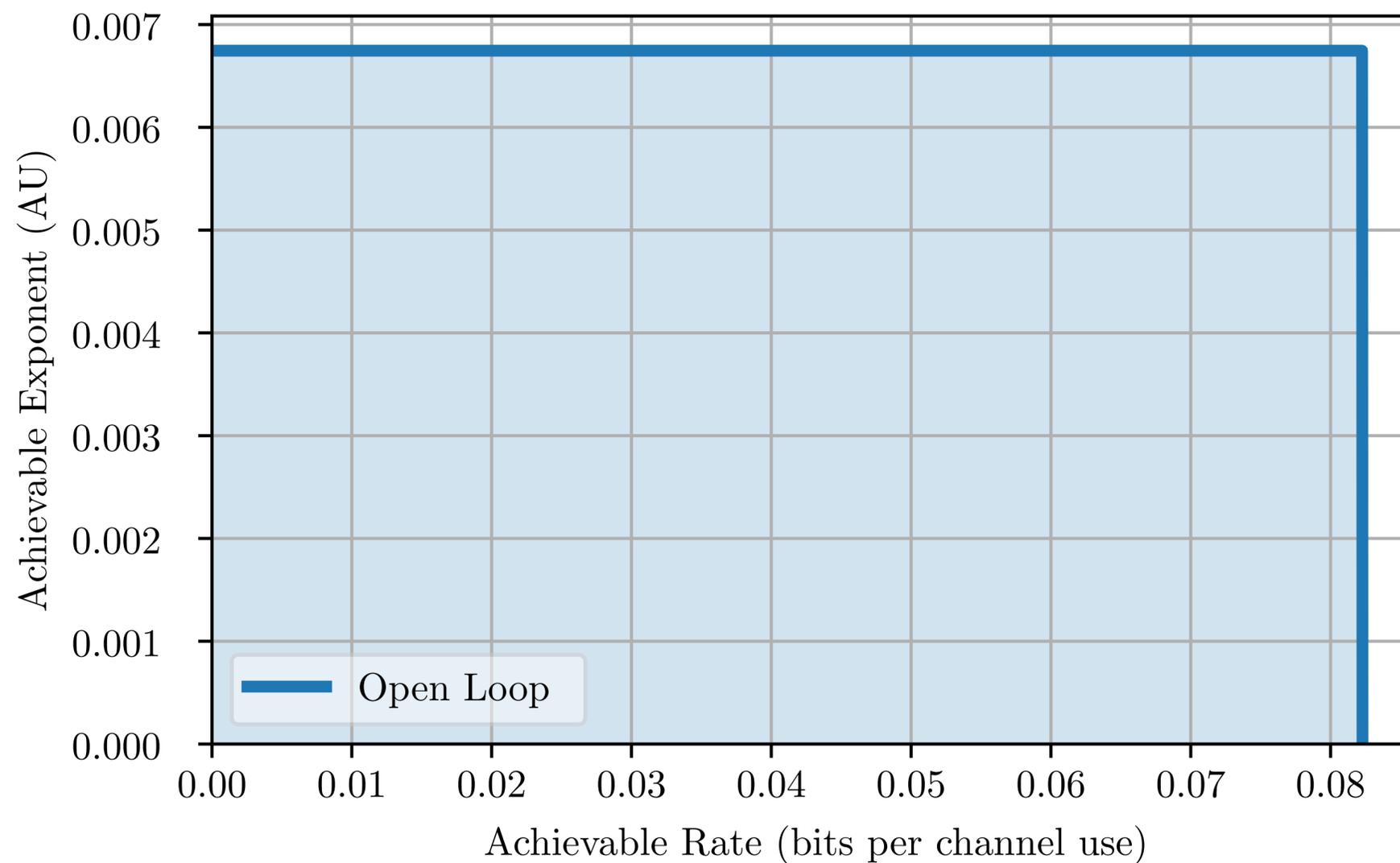
Table 1: Table for  $W_{Z|X,S}(0) = W_{Y|X,S}(0)$  for all  $X \in \{0, 1\}$  and  $S \in \{0, 1, 2\}$ .



## ► SIMULATION RESULT FOR NO TRADEOFF EXAMPLE

$S \backslash X$	0	1
0	0.9	0.1
1	0.8	0.2
2	0.7	0.3

Table 1: Table for  $W_{Z|X,S}(0) = W_{Y|X,S}(0)$  for all  $X \in \{0, 1\}$  and  $S \in \{0, 1, 2\}$ .



**DEFINITION: ACHIEVABILITY OF RATE AND EXPONENT PAIR**

$(R, E)$  is achievable if for every  $\epsilon > 0$ , there exists  $n$  large enough and a code  $\mathcal{C}$  of length  $n$  such that

$$\begin{aligned} P_c^{(n)} &\leq \epsilon, \\ E_d^{(n)} &\geq E - \epsilon, \\ \frac{1}{n} \log |\mathcal{C}| &\geq R - \epsilon \end{aligned}$$

▶ **SKETCH OF PROOF:**

- ▶ Fix the type of codewords,  $P_X$
- ▶ By the Compound Channel Coding Theorem [**Blackwell et al.'59**],  $R = \min_{s \in \mathcal{S}} \mathbb{I}(P_X, W_{Y|X,s})$  is achievable
- ▶ By Open Loop hypothesis testing [**Naghshvar-Javidi'13, Nitinawarat et al.'13**], the exponent

$$\phi(P_X) = \min_{s \in \mathcal{S}} \min_{s' \neq s} \max_{\ell \in [0,1]} - \sum_x P_X(x) \log \left( \sum_z W_{Z|x,s}(z)^\ell W_{Z|x,s'}(z)^{1-\ell} \right)$$

is asymptotically achievable

**SKETCH OF PROOF:**

- ▶ Assume that  $(R, E)$  is achievable
- ▶ Since there are at most polynomial number of types, there exists a type set  $\mathcal{T}$ , such that for all  $P_X \in \mathcal{T}$  with subcode  $\mathcal{C}_{P_X}$ 
  - ▶ Probability of error for reliable communication is small
  - ▶ Subcode includes almost all codes
  - ▶ Rate of the subcode is upper bounded by  $\min_s \mathbb{I}(P_X, W_{Y|X,s})$
- ▶ Choose  $P_X^* = \operatorname{argmin}_{P_X \in \mathcal{T}} \phi(P_X)$
- ▶ Then the probability of error is dominated by the probability of error of the subcode
  - ▶ Since it has a fixed type the error exponent is bounded by  $\phi(P_X^*)$

▶ **SKETCH OF PROOF:**

▶ We first define the three-phase encoder/decoder pairs:

- ▶ During the time  $t \in [1; \Delta_1 n]$  :
  - ▶  $P_X^\#$  is a type that maximizes  $\phi(P_X)$
  - ▶ Select a sequence  $\mathbf{v}$  that has the type  $P_X^\#$
  - ▶ Encoder maps any input message  $w \in [1; M]$  into  $\mathbf{v}$  and the state estimator is a ML detector.
  - ▶ This step is to estimate the state by observing the received sequence.
- ▶ During the time  $t \in [\Delta_1 n + 1; \Delta_2 n]$ 
  - ▶ Pick a channel code with size  $|\mathcal{S}|$  .
  - ▶ The encoder maps the estimated state from phase 1 into the corresponding codeword.
  - ▶ The channel code decoder at the receiver detect the state.
  - ▶ This step is to convey the information of the state to the receiver.
- ▶ During the time  $t \in [\Delta_2 n + 1; n]$ 
  - ▶ Pick a channel code with size  $M$  and has the type  $P_X$  ,  
 where the rate of the code is related to the estimated state in the phase 1.
  - ▶ The encoder maps the message  $w$  into the corresponding codeword.
  - ▶ The message decoder is the channel code decoder and the state estimator is the ML detector.

- ▶ Exact characterization of joint communication and sensing region for open loop as rate and detection-error exponent as off line resource allocation
- ▶ Lower bound characterization of the joint communication and sensing region for adaptive
- ▶ Clear benefits for using sensing for communication
- ▶ **FUTURE EXTENSIONS:**
  - ▶ More complicated model
  - ▶ Characterization of closed loop performance



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